# Data Communication Networks 

## Lecture 2

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## DLC

■ We are not going to study the physical layer and how communication signals are sent and received

- We will assume that we are capable of sending bits over a link
- DLC is responsible for reliable transmission of packets over a link
- every packet is delivered once,
- only once,
- without errors,
- and in order
- To achieve this goal, we have:
- Framing: determine start and end of packets
- Error detection: determine when errors exist
- Error correction: retransmit packets containing errors


## Framing

■ Recall that DLC adds its own header and trailer to the packet $\Rightarrow$ frame

$\underbrace{$|  header  |  packet  |  trailer  |
| :---: | :---: | :---: |}$_{\text {frame }}$

- The problem is to decide where successive frames start and end
- in some cases, there is a period of idle fills between successive frames (e.g. synchronous bit pipe)
- it is also necessary to separate idle fills from frames
- even when idle fill are replaces by dead periods (intermittent bit pipe), problem is not simplified, e.g. often no dead periods

$$
\text { ... } 01011010110010110101101110101110 \text {... }
$$

Where is the data?

## Character based framing

- Character based codes, such as ASCII (7 bits and 1 parity bit), provide binary representation for keyboard characters and terminal control characters
- Such codes can also provide representation for various communication characters
- SYN: a string of SYN characters provide idle fill between frames when a sending DLC has no data to send (but a synchronous modem requires bits)
- STX: start of text
- ETX: end of text

- Frame must contain integer number of characters
- Frame is character code dependent $\Rightarrow$ how do we send binary data
- e.g. packet is an arbitrary binary string and may possibly contain the ETX character for instance, which could be wrongly interpreted as end of frame


## Character based framing (cont.)

- A special control character DLE (Data Link Escape) is inserted before any intentional use of communication control characters
- e.g. DLE is not inserted before the possible appearance these characters as part of the binary data

But what if DLE appears itself in the data?

- Insert a DLE before each appearance of DLE in data
- e.g. DLE ETX (but not DLE DLE ETX): end of frame
- e.g. DLE DLE ETX (but not DLE DLE DLE ETX): DLE ETX in data

| SYN | SYN | DLE | STX | header | packet | DLE | ETX | CRC | SYN | SYN |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |

- Too much overhead: at least 6 characters/packet
- Primary framing method from 1960-1975


## Length field

- The basic problem is to inform the receiving DLC where each idle fill ends and where each frame ends
- idle fill: in principle, easy to identify because it is represented by a fixed string
- frame: harder to indicate where it ends because it consists of arbitrary and unknown bit string
- Include a length field of certain number of bits in header (e.g. DECNET)

- Once synchronized, DLC can always tell where next frame starts
- Length field restricts packet size
- length field must be $\left\lfloor\log _{2}\right.$ MaxFrameSize $\rfloor+1$ bits (that's the overhead)
- Difficult to recover from error in length field
- e.g. resynchronization needed after error in length field


## Maximum frame size

- How should transport layer choose maximum packet size?
- not a big deal since IP fragments packets further if necessary
- usually about 1500 bytes (from Ethernet)
- but theoretically speaking?
- Let $K_{\max }$ be the maximum packet size. Assume $V$ overhead bits. Let $M$ be the message length. Then we have

$$
M+\left\lceil\frac{M}{K_{\max }}\right\rceil V
$$

bits to send


Therefore,

- $K_{\max } \nearrow \Rightarrow$ small overhead per message
- $K_{\max } \searrow \Rightarrow$ faster delivery of message (why?)


## Maximum frame size (cont.)

- What is the time needed for the message to traverse $j$ links?

- Assume capacity of link is $c$ bps (bandwidth)

$$
\begin{gathered}
T=\frac{\left(K_{\max }+V\right)(j-1)}{c}+\frac{M+\left\lceil\frac{M}{K_{\max }}\right\rceil V}{c}+P+Q \\
E[T] \propto\left(K_{\max }+V\right)(j-1)+E[M]+\frac{E[M]}{K_{\max }} V
\end{gathered}
$$

- Minimizing (take first derivative and set it to zero)

$$
K_{\max }=\sqrt{\frac{E[M] V}{j-1}}
$$

## Fixed length packets/frames

- Length field is implicit (not needed)
- e.g. ATM, all packets are 53 bytes
- Requires synchronization upon initialization
- Message length not multiple of packet size
- last packet contains idle fill (efficiency?)


## Bit oriented framing

- In character based framing, DLE ETX indicates the end of frame
- avoided within frame by doubling each DLE character
- In bit oriented framing, a special binary flag indicates the end of frame
- avoided within frame using a technique called bit stuffing

■ The difference is that a flag can be of any length (later we see how to set length to minimize overhead)

- Standard protocols use 01111110, we denote it by $01^{6} 0$
- The same flag can be used to indicate start of frame

01111110 .01111110

- Standard DLCs have also an abort capability in which a frame can be aborted by sending 7 or more consecutive 1's (15 consecutive 1 's $\Rightarrow$ link is idle)
- Therefore, 0111111 , i.e. $01^{6}$ is the actual bit string that must be avoided in data

Bit stuffing, 1970 by IBM

## Bit stuffing

- Sender DLC
- insert (stuff) a 0 after each appearance of five consecutive 1 's
- append the flag $01^{6} 0$ (without stuffing) at the end of frame
- Receiver DLC
- delete the first 0 after each string of five consecutive 1's
- if six consecutive 1 's are seen $\Rightarrow$ end of frame


Which ones of these stuffed bits can be avoided (provided receiver's rule for deleting stuffed bits is changed accordingly)?

Qvevheas t tof diturtstuffing stuffing (assume flag $=01^{j} 0$ )?

- if $j \nearrow \Rightarrow$ less stuffing but longer flag
- if $j \searrow \Rightarrow$ more stuffing but shorter flag
- For the sake of analysis, assume a random string of bits with $p(0)=p(1)=1 / 2$
- insertion after $i^{t h}$ bit occurs with probability $(1 / 2)^{j}$

$$
0 \underbrace{1 \ldots 1}_{j-1} \downarrow
$$

- also insertion after $i^{t h}$ bit occurs with probability $(1 / 2)^{2 j-1}$

$$
0 \underbrace{1 \ldots 1}_{j-1} \underbrace{1 \ldots 1}_{j-1} \downarrow
$$

- since $(1 / 2)^{2 j-1} \ll(1 / 2)^{j}$, we can ignore such event and events of insertion dur to yet longer strings of 1 's
- The probability of stuffing after the $i^{t h}$ bit is approximately $2^{-j}$, which is also $E[$ stuffed @ $i]$
- By linearity of expectation, $E[s t u f f e d \mid K] \approx K 2^{-j}$ (be careful at boundary), where $K$ is the length of the frame


## Qverhead of bit stuffing (cont.)

$$
\begin{gathered}
E[\text { stuffed }]=E[K] 2^{-j} \\
E[\text { overhead }]=\underbrace{E[K] 2^{-j}}_{\text {stuffed }}+\underbrace{j+2}_{\text {flag }}
\end{gathered}
$$

- Minimizing with respect to $j$, we need smallest $j$ such that

$$
E[K] 2^{-j}+j+2<E[K] 2^{-(j+1)}+(j+1)+2
$$



$$
E[k] 2^{-(j+1)}<1
$$

■ The smallest $j$ that satisfies above is $j=\left\lfloor\log _{2} E[k]\right\rfloor$

- We can show that (homework?)

$$
\log _{2} E[K]+2.914 \leq E[\text { overhead }] \leq \log _{2} E[K]+3
$$

## Can we do better?

Length field based framing and bit oriented framing are comparable in their overhead

$$
\approx \log _{2} K
$$

where $K$ is the length of the frame

- Can we do better? Information theory tells us NO.
- Essentially, we are encoding information about the length of the frame at the sending DLC and transmitting it to the receiving DLC
- At least we need a number of bits equal to the entropy

$$
H=\sum_{k=1}^{K_{\max }} p(k) \log _{2} \frac{1}{p(k)}
$$

When distribution is uniform, i.e. $p=\frac{1}{K_{\max }}, H=\log _{2} K_{\text {max }}$.

## Error detection

All framing techniques are sensitive to errors

- error in DLE ETX
- error in length field (re-sync needed)
- error in flag
- flag ruined (frame disappears)
- flag created by error (extra frame appears)
- error in data itself
- Flag approach is least sensitive to errors because a flag will eventually appear
- the only thing is that an erroneous packet/frame is created
- but this can be removed by error detection techniques
- Error detection is used by receiving DLC to determine if a frame contains errors
- If frame contains errors, receiver requires the transmitter to resend the frame


## How to detect errors?

- The problem (simply stated)
- Assume that DLC knows where frames begin and end (solved earlier)
- Determine which frames contain errors
- Error cannot be detected by analyzing the packet itself (why?)
- Therefore, extra bits must be used
- Parity check
- single parity
- multiple parity (e.g. horizontal and vertical)
- Cyclic Redundancy Check CRC


## Single parity check

- Add one parity bit
- Parity bit is 1 if frame contains ODD number of 1 's, and 0 otherwise
- e.g. $1 0 0 1 0 1 0 \longdiv { 1 }$
- e.g. 01110100
- Therefore, a frame always contains an even number of 1 's
- Receiver counts number of 1 's
- ODD number of 1's: an error must have occurred
- EVEN number of 1's: interpret as no error (why?)
- even number of errors cannot be detected!
- $p($ undetected error $)=\sum_{i}$ even $\binom{k}{i} p^{i}(1-p)^{i}$ assuming independent errors (simplification), where $k$ is the length of the frame, $p$ is the probability of error (binary symmetric channel)


## Horizontal and vertical parity checks

Data is visualized as a rectangular array

| 1 | 0 | 0 | 1 | 0 | 1 | 0 |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 1 | 1 | 1 | 0 | 1 | 0 |
| 1 | 1 | 1 | 0 | 0 | 0 | 1 |
| 1 | 0 | 0 | 0 | 1 | 1 | 1 |
| 0 | 0 | 1 | 1 | 0 | 0 | 1 |

- Parity bit is computed for every row and every column
- If an even number of errors is confined to a single row, each of them can be detected by the corresponding column parity checks (and vice-versa)


## Horizontal and vertical parity checks

- Data is visualized as a rectangular array

| 1 | 0 | 0 | 1 | 0 | 1 | 0 | 1 |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 1 | 1 | 1 | 0 | 1 | 0 | 0 |  |
| 1 | 1 | 1 | 0 | 0 | 0 | 1 | 0 | horizontal checks |
| 1 | 0 | 0 | 0 | 1 | 1 | 1 | 0 |  |
| 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 |  |
| 1 | 0 | 1 | 1 | 1 | 1 | 1 | 0 | always consistent with <br> (why?) <br> (addition modulo 2) |

vertical checks

- Parity bit is computed for every row and every column
- If an even number of errors is confined to a single row, each of them can be detected by the corresponding column parity checks (and vice-versa)


## Horizontal and vertical parity checks

Data is visualized as a rectangular array

| 1 | 0 | 0 | 1 | 0 | 1 | 0 | 1 |  |
| :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- | :--- |
| 0 | 1 | 1 | 1 | 0 | 1 | 0 | 0 |  |
| 1 | 1 | 1 | 0 | 0 | 0 | 1 | 0 | horizontal checks |
| 1 | 0 | 0 | 0 | 1 | 1 | 1 | 0 |  |
| 0 | 0 | 1 | 1 | 0 | 0 | 1 | 1 |  |
| 1 | 0 | 1 | 1 | 1 | 1 | 1 | 0 | halways consistent with both checks <br> (why?) (addition modulo 2) |
| ald |  |  |  |  |  |  |  |  |

## vertical checks

- Parity bit is computed for every row and every column
- If an even number of errors is confined to a single row, each of them can be detected by the corresponding column parity checks (and vice-versa)
- Some errors are still undetected
- e.g. any 4 errors forming a rectangle


## Arbitrary parity check codes

Parity check code is simply addition modulo 2

$$
\underbrace{s_{1} s_{2} \ldots s_{k}}_{K \text { bit frame }} \underbrace{c_{1} c_{2} \ldots c_{L}}_{L \text { bit parity check }}
$$

- every $c_{i}$ is the sum of some bits in $s_{1} \ldots s_{K}$

$$
c_{i}=\sum_{j=1}^{K} \alpha_{i j} s_{j}
$$

where $\underset{S_{1}}{\alpha}$ is ${\underset{S}{2}}^{\text {an }} L \underset{S_{3}}{\times} K \quad{ }_{C_{1}}^{0-1}{\underset{C}{C}}^{2}$ matrix

|  | ${ }_{2}$ | ${ }_{3}{ }_{3}$ |  | $\mathrm{max}^{\text {c }}$ |  | $c_{4}$ |  |  |  |
| :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: | :---: |
| 0 | 0 | 0 | 0 | 0 | 0 | 0 |  |  |  |
| 0 | 0 | 1 | 1 | 1 | 0 | 1 |  |  |  |
| 0 | 1 | 0 | 0 | 1 | 1 | 1 | $c_{1}=s_{1}+s_{3}$ |  | $\left[\begin{array}{lll}1 & 0 & 1 \\ 1 & 1 & 1\end{array}\right.$ |
| 0 | 1 | 1 | 1 | 0 | 1 | 0 | $c_{2}=s_{1}+s_{2}+s_{3}$ |  | $\begin{array}{lll}1 & 1 & 1 \\ 1 & 1 & 0\end{array}$ |
| 1 | 0 | 0 | 1 | 1 | 1 | 0 | $c_{3}=s_{1}+s_{2}$ |  | $\begin{array}{lll}1 & 1 & 0 \\ 0 & 1 & 1\end{array}$ |
| 1 | 0 | 1 | 0 | 0 | 1 | 1 | $c_{4}=s_{2}+s_{3}$ |  |  |
| 1 | 1 | 0 | 1 | 0 | 0 | 1 |  |  |  |
| 1 | 1 | 1 | 0 | 1 | 0 | 0 |  |  |  |

## Effectiveness of a code

$\underbrace{$| K | L |
| :---: | :---: |}$_{\text {codeword }}$

The effectiveness of a code is usually measured by three parameters
■ minimum distance of the code $d$ : the smallest number of errors that can convert one code word into another

- burst detecting capability
- burst = number of bits from first error to last error (inclusive)
- defined as: largest integer $B$ such that a code can detect all bursts $\leq B$
- probability that a random string is accepted as error free
- useful when framing is lost, e.g. check code is random with respect to received frame
- We have $2^{K}$ codewords (why?) and $2^{K+L}$ random strings
- therefore, the probability is $2^{-L}$


## Effectiveness of a code (cont.)

- Minimum distance $d$
- single parity:
- horz. and vert. parity:
- Burst detecting capability $B$
- single parity:
- horz. and vert. parity (assumes data sent by rows):


## Cyclic Redundancy Check

For convenience, denote the data bits as

$$
s_{K-1}, s_{K-2}, \ldots, s_{0}
$$

Represent the string as a polynomial

$$
s(x)=s_{K-1} x^{K-1}+s_{K-2} x^{K-2}+\ldots+s_{1} x+s_{0}
$$

- Similarly, we can represent the CRC (with $L$ bits) as

$$
c(x)=c_{L-1} x^{L-1}+c_{L-2} x^{L-2}+\ldots+c_{1} x+c_{0}
$$

- The whole frame can be represented as a polynomial

$$
f(x)=s(x) x^{L}+c(x)=\underbrace{s_{K-1}} x^{L+K-1}+\ldots+\underbrace{s_{0}} x^{L}+\underbrace{c_{L-1}} x^{L-1}+\ldots+\underbrace{c_{0}}
$$

- Why this polynomial representation? because we are going to obtain $c$ as $c(x)$ by dividing $s(x) x^{L}$ by some polynomial $g(x)$


## Obtaining $c(x)$

■ We know $s_{i}$ (data) for $i=0 \ldots K-1$

- How do we compute $c_{i}(\mathrm{CRC})$ for $i=0 \ldots L-1$ ?
- let $g(x)=x^{L}+g_{L-1} x^{L-1}+\ldots+g_{1} x+1$ be given $\left(g_{L}=g_{0}=1\right)$
- then

$$
c(x)=\text { Remainder }\left[\frac{s(x) x^{L}}{g(x)}\right]
$$

$\nwarrow$ division modulo 2

- result is a degree $L-1$ polynomial $\Rightarrow L$ bits
- Example: $s=101(K=3)$ and $g(x)=x^{3}+x^{2}+1(L=3)$
- $s(x)=$ ?
- $s(x) x^{L}=$ ?
- divide $s(x) x^{L}$ by $g(x)$ and obtain remainder (i'll do it on the board?)


## Another example

$s=110101$
$g(x)=x^{3}+1$
$s(x)=x^{5}+x^{4}+x^{2}+1$
$s(x) x^{L}=x^{8}+x^{7}+x^{5}+x^{3}$

$$
\begin{array}{l|l}
x^{8}+x^{7}+x^{5}+x^{3} & \frac{x^{3}+1}{x^{5}+x^{4}+x+1} \\
\frac{x^{8}+x^{5}}{x^{7}+x^{3}} & \\
\frac{x^{7}+x^{4}}{x^{4}+x^{3}} & \\
\frac{x^{4}+x}{x^{3}+x} & \\
\frac{x^{3}+1}{x+1} &
\end{array}
$$

$c(x)=0 \cdot x^{2}+1 \cdot x+1(L=3) \Rightarrow c=011$

## Using bits only

$$
\begin{aligned}
& s=110101 \\
& g(x)=x^{3}+1(L=3) \\
& g=\begin{array}{cccccccccc} 
\\
& 1 & 0 & 0 & 1 & & & & & \\
\\
& \downarrow & \downarrow & \downarrow & \downarrow & & & & & \\
\\
x^{3} & x^{2} & x^{1} & x^{0} & & & & & & \\
\\
& & & & & & & & & \\
\\
s(x) x^{L} \Rightarrow & 1 & 1 & 0 & 1 & 0 & 1 & 0 & 0 & 0 \\
& \downarrow & \downarrow & \downarrow & \downarrow & \downarrow & \downarrow & \downarrow & \downarrow & \downarrow \\
& x^{8} & x^{7} & x^{6} & x^{5} & x^{4} & x^{3} & x^{2} & x^{1} & x^{0}
\end{array}
\end{aligned}
$$

## Using bits only (cont.)



## Using bits only (cont.)



- Multiply $g$ by first bit
- Add
- Shift
- Can be implemented using a feedback shift register


## Feedback shift register



- Register is initialized with first $L$ bits of $s$
- After $K$ shifts, switch is moved and CRC is read

How does $c(x)$ help?

$$
\begin{gathered}
s(x) x^{L}=g(x) z(x)+c(x) \\
s(x) x^{L}+c(x)=g(x) z(x)+\underbrace{c(x)+c(x)}_{0} \text { (modulo 2) } \\
s(x) x^{L}+c(x)=g(x) z(x) \\
f(x)=g(x) z(x)
\end{gathered}
$$

- Polynomial representation of the frame is multiple of $g(x)$
- Assume $f(x)$ received as $y(x)$
- Receiver DLC computers

$$
\text { Remainder }\left[\frac{y(x)}{g(x)}\right]
$$

- if remainder is not zero $\Rightarrow$ error in frame
- if remainder is zero, declare the frame error free


## Undetected errors

- Assume error is $e(x)$, i.e. $y(x)=f(x)+e(x)$

■ Then, $\frac{y(x)}{g(x)}=\frac{f(x)}{g(x)}+\frac{e(x)}{g(x)}$

- Therefore, we have undetected errors iff $e(x) \neq 0$ divisible by $g(x)$
- Single errors are always detected
- assume undetected, i.e. $e(x)=x^{i}=g(x) z(x)$ for some $i$
- since $g(x)=x^{L}+\ldots+1$, multiplying $g(x)$ by any $z(x) \neq 0$ cannot produce $x^{i}$ (must produce at least 2 terms)
- $g(x)$ can be chosen such that
- all odd number of errors are detected
- all double errors are detected (if $K+L<2^{L}$ )
- therefore, minimum distance $d=4$
- burst detecting capability $B=L$
- probability of random string accepted is $2^{-L}$
- e.g. $(\mathrm{L}=16) g(x)=x^{16}+x^{15}+x^{2}+1$ CRC-16
- e.g. $(\mathrm{L}=16) g(x)=x^{16}+x^{12}+x^{5}+1$ CRC-CCITT
- e.g. $(\mathrm{L}=32) g(x)=\ldots$ (see book page 64$)$

